

Ballistic to Diffusive Crossover in III–V Nanowire Transistors

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Abstract—In this paper, we examine the crossover between ballistic and diffusive transport in III–V nanowire transistors. We find that at lower drain voltages the ballistic-to-diffusive crossover occurs at channel lengths of approximately 2.3 nm at room temperature. However, when we increase the drain voltage, we find that the ballistic-to-diffusive crossover can be more than nine times as long at room temperature. As the temperature is decreased, we find that the initially the performance is not significantly improved. However, as the temperature approaches 100 K, the ballistic-to-diffusive crossover increases to longer channel lengths quite dramatically. When InAs, InSb, and InP nanowires are compared at room temperature, we find that InAs and InSb perform in a similar fashion each with ballistic regions in excess of 10 nm, but that InP has no significant ballistic regime. Finally, we simulate several 10-nm InAs trigate transistors and show that for dopants deeply buried in the source and drain the devices appear ballistic, but when dopants appear near the source-channel interface, significant reductions in performance occur.

Index Terms—Nanowire transistors (NWTs), phonon interactions, quantum simulation, III–V.

I. INTRODUCTION

THE NANOWIRE transistor (NWT) [1] may be a viable device to replace bulk Si MOSFETs due to its ability to reduce short-channel effects and give greater control over the channel electron density. There has also been a resurgence in the exploration of III–V materials for next generation CMOS technology. While silicon is the unquestioned material of choice in CMOS technology, III–V materials have some advantages over silicon such as higher mobility, which could potentially lead to very high-speed low-power transistors. Of course, we recognize the tradeoff in terms of low density-of-states, i.e., Landauer channels. InSb has measured electron mobilities of $7.8 \times 10^4 \text{ cm}^2 \text{ V}^{-1} \cdot \text{s}^{-1}$ [2] and has long been touted as a good candidate for high-frequency devices. Recently, high-mobility wrap-gated p-channel InAs NWTs have been experimentally demonstrated which also have exceptionally high mobilities and solid device performance [3]. With the advent of technology capable of producing high quality III–V nanowires for transistor applications, it is natural to question the limits of device performance in III–V materials.

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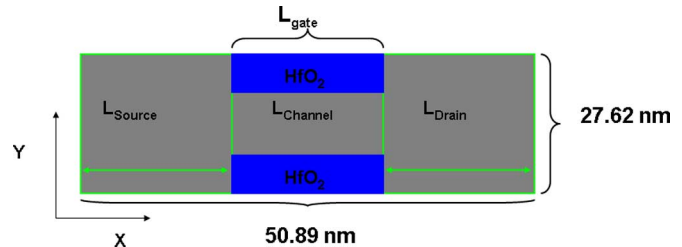


Fig. 1. Schematic of the device used in this paper taken in the xy plane.

Of particular interest for future generation CMOS applications is the location of the ballistic-to-diffusive crossover in III–V nanowire systems.

There have been many previous studies demonstrating the performance of silicon NWT [4]–[6] and InAs NWT [7]. However, since we are interested in operating these devices at or near room temperature, phonon processes are very important. A recent study has placed the ballistic-to-diffusive crossover for silicon nanowire systems at 1.42 nm [8]. This places serious limits on the effectiveness of silicon in next generation CMOS but it also shows the importance of phonon modes in the determination of the performance characteristics. In this paper, we examine the dependence of the ballistic-to-diffusive (mobility dominated transport) crossover point in III–V NWT as a function of temperature, drain voltage, and semiconductor material covering three of the most popular III–V materials: InAs, InSb, and InP using a fully 3-D self-consistent quantum transport formalism. The effects of phonon scattering are included in the simulations as separable self-energy terms in the Hamiltonian. To make the simulations as complete as possible, we include interactions with the dopant ions, and inelastic effects (the effects of acoustic deformation potential, intervalley phonons [$\Gamma \rightarrow X$ and $\Gamma \rightarrow L$], and polar optical phonons (POPs)].

II. DEVICE STRUCTURE AND SIMULATION METHOD

In Fig. 1, we display a schematic of the system under consideration in the x – y plane (the z -axis is normal to the plane shown). The device dimensions (in multiples of the lattice constant) use in the simulations were chosen to aid in the inclusion of the discrete dopants. The thickness of the InAs layer is 9.09 nm and the width of the channel is 8.48 nm. The source and drain of the device are n -type with a doping density of $1.3 \times 10^{18} \text{ cm}^{-3}$, while the channel of the device is considered to be p -type, but undoped. The gate material is assumed to be platinum and the gate oxide on each side is 1 nm of hafnium oxide (HfO_2). Underneath the device, we

79 have assumed a thick silicon dioxide substrate. The length of
80 the device is set to 50.89 nm. While the channel of the device
81 is typically much smaller than the overall length of the device
82 geometry, a large domain is set so that as the length of the
83 channel is increased, the bandstructure for the overall device
84 remains constant. The length of the channel starts at a length
85 of 6.0583 Å and is increased by this length until the diffusive
86 regime is obtained. During this increase of the length, the width,
87 and depth of the nanowire channel are kept constant. The length
88 of the gate that surrounds the nanowire on three sides is set to
89 be the same length as the channel of the device. The source and
90 drain of the device are divided equally depending on the length
91 of the channel; however, they never are reduced below 10 nm
92 in the x -direction.

93 Once the device geometry is defined, the InAs lattice is
94 scanned and the dopants are distributed according to the method
95 presented in [9]. Following the distribution of the dopants, they
96 are then mapped back onto the grid of the simulation mesh and
97 the initial self-consistent Poisson solution is obtained. In this
98 case, the full Coulomb potential of the dopants is incorporated.
99 Then, the solution of Poisson's equation for the local potential
100 is no longer smoothly varying in the source and drain of the
101 device. The inclusion of discrete dopants causes the formation
102 of potential variations in the source and drain. The density
103 throughout the device is calculated using a variant of the re-
104 cursive scattering matrix method which solves the Schrödinger
105 equation in the effective mass approximation, as described in
106 [4]. In short, this method provides us with a 3-D, fully quantum
107 mechanical treatment. Since the transport calculation is per-
108 formed in real space, the different excited modes in the system
109 are automatically coupled unlike in other simulations [6]. In
110 order to achieve self-consistency, the density obtained from the
111 transport calculation is then updated using Broyden's method
112 [10] and a new guess for the potential is obtained through
113 the solution of Poisson's equation. The process is repeated
114 until a desired level of convergence is obtained. Exchange and
115 correlations terms are included in each simulation through a
116 local density approximation [11].

117 In this paper, we seek the ballistic-to-diffusive crossover in
118 these nanowire systems. Our procedure for determining this
119 quantity is given below. We know that even when the transport
120 is ballistic, the lower limit of the resistance of the channel is
121 determined by the inverse of the Landauer conductance

$$R_{\text{ballistic}} = \left[\frac{2e^2}{h} N \right]^{-1} \quad (1)$$

122 where N is the number of transverse modes flowing in the
123 quantum wire. Clearly, the resistance in (1) has no dependence
124 on the length of the quantum wire. Conversely, when the
125 transport is determined by the carrier mobility and the carrier
126 density, then the resistance becomes

$$R_{\text{diffusive}} = \frac{1}{ne\mu} \frac{L}{A} \quad (2)$$

127 where L is the channel length and A is the cross-sectional
128 area of the inversion layer. We have been careful to use the

area of the inversion layer rather than the width because we
129 are dealing with a 3-D quantum wire with quantization in the
130 transverse directions. In this regime, there is a clear linear de-
131 pendence of the resistance with the length of the quantum wire
132 channel. 133

III. POP SCATTERING

134

The inclusion of separable scattering mechanisms is not a
135 new concept. In previous work, acoustic and optical phonon
136 processes have been derived [8] and so the work will not be
137 repeated here. However, we are now interested in including the
138 nonlocal effects of POPs which, to this point, have not been
139 treated. This is a crucial difference between silicon and III-V
140 materials. In silicon, the local phonon processes dominate,
141 but in III-V materials nonlocal optical processes dominate the
142 room temperature transport. We begin by noting that to use the
143 method presented in [8], we are assuming that the scattering
144 is weak relative to the energies in the system, and that we
145 are only interested in the steady-state limit. Therefore, we can
146 use the Fermi golden rule expression, equivalent to a first-
147 order nonself-consistent Born approximation, for each of the
148 scattering processes and generate a real space self-energy from
149 it. In particular, we note that the imaginary part of the self-
150 energy term is related to the scattering rate and it is the latter
151 scattering rate that we wish to calculate [8]. This will result in
152 an x -directed momentum which is related to the carrier energy
153 in the quantum wire. While the local scattering rates have been
154 previously derived [8], we will present the important changes
155 in the general derivation approach to take into account the
156 nonlocal nature of POP scattering. In this paper, we use the
157 Fermi golden rule to calculate scattering rates. This is treated in
158 many other places, however, we must account for the transverse
159 modes of the quantum wire. Therefore, we begin with the
160 general form 161

$$\begin{aligned} \left(\frac{1}{\tau_{\text{polar}}} \right)_{ij}^{i'j'} &= \frac{\pi e^2}{\gamma V \omega_0} \sum_{q_x} \sum_{q_y, q_z} \int dy \int dz \int dy' \int dz' \\ &\times \varphi_{i,j}^*(y, z) \varphi_{i',j'}(y, z) \varphi_{i,j}^*(y', z') \varphi_{i',j'}(y', z') \\ &\times \frac{e^{iq_y(y-y') + iq_z(z-z')}}{q^2} \\ &\times \{ (N_q + 1) \delta(E_k - E_{k-q} - \hbar\omega_0) \\ &\quad + N_q \delta(E_k - E_{k-q} + \hbar\omega_0) \} \quad (3) \end{aligned}$$

where V is the volume, ω_0 is the phonon frequency and γ is
162 the effective interaction parameter, i is the initial mode index
163 in the width direction, j is the initial mode index in the depth
164 direction, i' is the final mode index in the width direction,
165 and j' is the final mode index in depth direction. The delta
166 functions in (3) serve to conserve the energy in the process
167 of the interactions of the carriers with the POPs. $\varphi_{i,j}(y, z)$ is
168 the transverse 2-D wave function in the particular slice under
169 consideration. We assume in this treatment that we are dealing
170 with parabolic bandstructure. Taking a closer look at the terms 171

172 contained within the delta functions, we realize that based on
173 this assumption

$$\begin{aligned} E_k - E_{k\pm q} \pm \hbar\omega_0 &= E_{i,j} - E_{i',j'} \pm \hbar\omega_0 + \frac{\hbar^2 k_x^2}{2m_x^*} - \frac{\hbar^2 (k_x \pm q_x)^2}{2m_x^*} \\ &= E_{i,j} - E_{i',j'} \pm \hbar\omega_0 - \frac{\hbar^2 k_x q_x}{m_x^*} \pm \frac{\hbar^2 q_x^2}{2m_x^*}. \end{aligned} \quad (4)$$

174 In (4), $E_{i,j}$ and $E_{i',j'}$ are the initial and final energies at the
175 bottom of their respective bands. We begin to simplify (3), by
176 first examining the longitudinal terms

$$\sum_{q_x} \frac{1}{2q_x^2} \delta(E_k - E_{k\pm q} \pm \hbar\omega_0). \quad (5)$$

177 We now expand (5) using the results from (4)

$$\sum_{q_x} \frac{1}{q_x^2 + q_y^2 + q_z^2} \delta\left(\mp \frac{\hbar^2 q_x^2}{2m_x^*} \mp \frac{\hbar^2 k_x q_x}{m_x^*} + E_{i,j} - E_{i',j'} \pm \hbar\omega_0\right). \quad (6)$$

178 This sum may be simplified further by taking the Fourier
179 transform with respect to q_x

$$\frac{1}{2\pi} \int_{-\infty}^{\infty} \frac{dq_x e^{iq_x(x-x')}}{q_x^2 + q_y^2 + q_z^2} \delta\left(\mp \frac{\hbar^2 q_x^2}{2m_x^*} \mp \frac{\hbar^2 k_x q_x}{m_x^*} + \Delta_{i',j'}\right) \quad (7)$$

180 where we denote

$$\Delta_{i',j'} = E_{i,j} - E_{i',j'} \pm \hbar\omega_0. \quad (8)$$

181 Representing (7) as a summation, we have

$$\begin{aligned} &\frac{1}{2\pi} \int_{-\infty}^{\infty} \frac{dq_x e^{iq_x(x-x')}}{q_x^2 + q_y^2 + q_z^2} \frac{1}{|\partial E_{k\pm q}/\partial q_x|} \\ &\frac{m_x^*}{2\pi\hbar^2} \int_{-\infty}^{\infty} \frac{dq_x e^{iq_x(x-x')}}{q_x^2 + q_y^2 + q_z^2} \frac{1}{\left|\mp k_x + \frac{q_x}{2}\right|}. \end{aligned} \quad (9)$$

182 At this point, we examine the relationship for $E_{k\pm q}$

$$E_{k\pm q} = \frac{\hbar^2 k_x^2}{2m_x^*} \pm \frac{\hbar^2 k_x q_x}{m_x^*} + \frac{\hbar^2 q_x^2}{2m_x^*}. \quad (10)$$

183 Equation (10) is quadratic in q_x and may then be solved to yield
184 a solution for q_x which is then substituted back into (9) making
185 the term in the absolute value independent of q_x as shown
186 below regardless of if the phonon process under consideration
187 is emission or absorption. Therefore, we may group all of
188 the variables that are independent of q_x into a simple dummy
189 variable a . With this pairing, we may reduce our summation in
190 (9) to reach

$$\frac{m_x^*}{2\pi\hbar^2} \int_{-\infty}^{\infty} \frac{dq_x e^{iq_x(x-x')}}{q_x^2 + a^2}. \quad (11)$$

Equation (11) is most easily evaluated by making a transforma-
tion to cylindrical coordinates to obtain

$$\frac{m_x^*}{4\pi\hbar^2} \iint \frac{q_x e^{iq_x(\vec{r} - \vec{r}')}}{q_{xi}^2 + a^2} d\theta dq. \quad (12)$$

We evaluate the integral in (12) using contour integrations to
arrive at a final, simplified expression for the sum over the
longitudinal wavevectors

$$\sum_{q_x} \frac{1}{2q_x} \delta(E_k). \quad (13)$$

The expression in (13) is substituted back into (3) to find an
overlap integral that is similar to both the one found in [8] and
in most textbooks

$$\begin{aligned} &= \sum_{q_x} \sum_{q_y, q_z} \int dy \int dz \int dy' \int dz' \varphi_{i,j}^*(y, z) \varphi_{i',j'}(y, z) \\ &\quad \times \varphi_{i,j}^*(y', z') \varphi_{i',j'}(y', z') \frac{e^{iq_y(y-y')} + iq_z(z-z')}{2q_x} \delta(E_k) \\ &= \frac{A}{2V} \sum_{q_x} \iint dz dy [\varphi_{i,j}^*(y, z) \varphi_{i',j'}(y, z)]^2 \frac{1}{q_x} \delta(E_k) \\ &= \sum_{k_x} I_{i,j}^{i',j'} \frac{1}{k_x} \delta(E_k). \end{aligned} \quad (14)$$

Where in (14), we have used the following definition

$$I_{i,j}^{i',j'} = \frac{A}{4\pi V} \iint dy dz [\varphi_{i,j}^*(y, z) \varphi_{i',j'}(y, z)]^2. \quad (15)$$

Following the usual procedure, we replace the summation over
the final momentum states with an integration

$$\sum_{k'} \rightarrow \frac{L}{2\pi} \int_{-\infty}^{\infty} dk' = \int_0^{\infty} \rho_{1D}(E') dE' \quad (16)$$

to obtain the final result

$$\left(\frac{1}{\tau_{\text{polar}}}\right)_{i,j}^{i',j'} = \frac{m_x^* L e^2}{32\pi^3 \gamma \omega_0} I_{i,j}^{i',j'} \int \frac{dk_x}{k_x} \delta(E_k). \quad (17)$$

However, we are still in the mode space representation and
the longitudinal momentum is not yet a viable operator. From
many-body physics, the momentum dependence arises from the
Fourier transform of the differences in the two coordinates in
the self-energy. Therefore, to include the proper momentum
dependence, we take the inverse transform. This essentially
involves solving a contour integral for both the emission and
absorption cases which results from (17). This gives us our
final, simplified and summarized results for absorption and
emission of POPs

$$\left(\frac{1}{\tau_{\text{polar}}}\right)_{i,j}^{i',j'} = \frac{m_x^* L e^2}{32\pi^3 \gamma \omega_0} I_{i,j}^{i',j'} \begin{cases} N_q e^{-k_0|x-x'|} \\ (N_q + 1) \frac{\sin(k_0|x-x'|)}{k_0|x-x'|} \end{cases}. \quad (18)$$

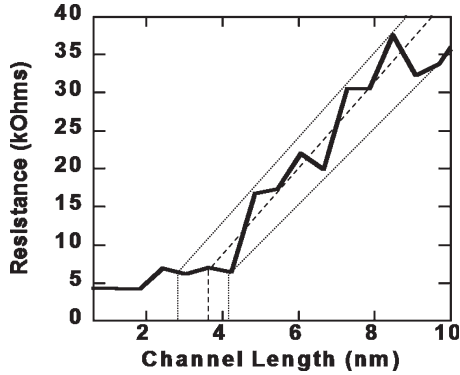


Fig. 2. Resistance of an InAs quantum wire as a function of the length of the channel. The gate voltage has been set to $V_g = 0.5$ V and the drain voltage has been set to $V_d = 10$ mV. The temperature of the device is 300 K. The red lines signify the maximum and minimum ranges for the ballistic to diffusive crossover, and the blue line signifies the median value. These lines are used to determine the value quoted for the ballistic to diffusive crossover.

213 This result is reasonable as we now see the manifestation of
 214 the nonlocal nature of POP scattering as a clear dependence on
 215 distance in (18). To utilize this in the transport calculation, we
 216 must use the following unitary transformation to convert this
 217 form to the site representation for inclusion in the Hamiltonian

$$\Gamma_{\text{polar}} = \text{Im}\{\Sigma\} = U^+ \left(\frac{\hbar}{\tau_{\text{polar}}} \right)_{i,j}^{i',j'} U \quad (19)$$

218 where U is a unitary mode-to-site transformation matrix. The
 219 unitary matrix U^+ results from the eigenvalue solutions in the
 220 transverse slice and are composed of the various eigenfunc-
 221 tions in the site basis. Hence, it represents a mode-to-slice
 222 transformation.

223 IV. RESULTS

224 A. Drain-Voltage Dependence

225 In Fig. 2, we plot the resistance of an InAs quantum wire
 226 as a function of the length of the channel for a small drain
 227 bias of $V_d = 10$ mV. In each of the following plots, the gate
 228 voltage is held constant at $V_g = 0.5$ V in order to ensure that
 229 there is a channel in the p -type wire. It should be noted that the
 230 optical phonon energy used in the calculation of InAs devices
 231 is 30 meV. While the channel length remains below 3.6 nm, the
 232 resistance of the device does not appear to change significantly.
 233 The oscillations in the resistance in Fig. 2 arise mainly from
 234 the interaction of the incident electron modes with the material
 235 boundaries, and in particular, with the source and channel
 236 interfaces creating strong quantum interference which is seen
 237 not only in theoretical studies [12], [13] but also experimentally
 238 [14]. For channels longer than about 3.6 nm, the resistance
 239 shows a clear linear dependence as the channel length increases
 240 as is expected when the nature of the transport changes from
 241 ballistic to diffusive. At this drain voltage, we estimate the
 242 ballistic-to-diffusive crossover to occur at 3.12 ± 0.3 nm. This
 243 is limited by the strength of the POP scattering as the X and
 244 L valleys are offset by sufficient energy to significantly limit

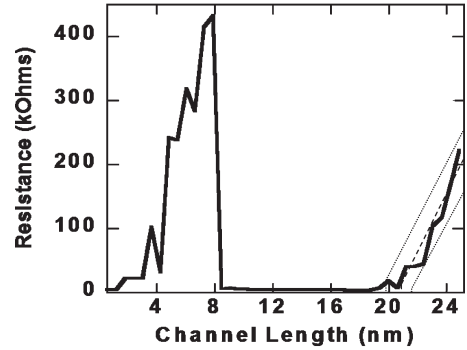


Fig. 3. Resistance of an InAs gated quantum wire as a function of the length of the channel. The gate voltage has been set to $V_g = 0.5$ V and the drain voltage has been set to $V_d = 0.5$ V. The temperature of the device is 300 K. The red lines once again signify the maximum and minimum ranges for the ballistic to diffusive crossover, and the blue line signifies the median value.

the intervalley scattering rates to these valleys. While this value
 245 is more than that of the quoted value of 1.42 nm for similar
 246 cross-sectional silicon quantum wires, this is not the significant
 247 increase we were hoping to see. Further, there will not be
 248 significant device-to-device variation in InAs or other III-V
 249 NWTs due to low concentration of dopant atoms as compared
 250 to that of a typical silicon device. 251

In Fig. 3, we now increase the drain voltage to $V_d = 0.5$ V.
 252 We find that this drain voltage consists of much more structure
 253 than the device held at reduced drain voltages. We begin our
 254 examination of Fig. 3, by noticing that there is an initial period
 255 of slightly more than 3 nm where the resistance of the quantum
 256 wire channel does not change. In this region, the length of the
 257 channel is such that the incident electrons see almost no channel
 258 potential and simply tunnel from the source to the drain of
 259 the device. As the length of the channel is increased beyond
 260 4 nm, we see that there is a large spike in the resistance of the
 261 channel. This behavior is attributed to the fact that the locations
 262 of the longitudinal states in the channel are too high in energy
 263 to support any modes and the channel becomes a large tunnel
 264 barrier. This argument is supported by the fact that the form of
 265 the drain voltage versus channel length in the regime of 4 to
 266 8 nm has an exponential form which is indicative of tunneling
 267 transport. We do not see this in the lower drain voltage sweeps
 268 due to the fact that scattering has blurred energy spectrum of
 269 the electrons before they reach the channel. This widening of
 270 the range of energies allows electrons not energetically capable
 271 of accessing the states to scatter into them. Thus, we see no rise
 272 in resistance in Fig. 2. While this type of interaction is fairly
 273 uncommon in semiconductor devices due to their relatively
 274 large sizes, this is common in systems of quasi-1-D electrons
 275 connected to external reservoirs [13]. 276

The existence of the tunneling regime in Fig. 3 is an im-
 277 portant point. If indeed, the physics is as we state then the
 278 same tunneling regime should show up in simple ballistic
 279 simulations of the same device structure. In Fig. 4, we plot
 280 the resistance of the quantum wire channel as a function of its
 281 length with no scattering included in the simulation. We find a
 282 similar trend as in the case with the inelastic processes included.
 283 In this simulation, we initially begin with three propagating
 284 modes in the system. The initial resistances do not match the 285

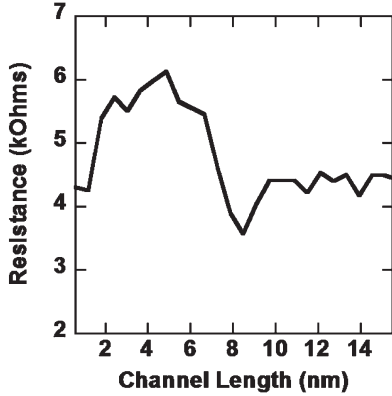


Fig. 4. Plot of the resistance of the quantum wire as a function of its length with no scattering included in the simulation for $V_g = V_d = 0.5$ V. We find the same short channel resistance trend in the ballistic case as we find in the case with the phonon processes included. Since the resistance increase in the quantum wires is the focus, we plot only the first 8 nm of channel length.

286 other quantum wires as doping is slightly lower than in other
 287 devices. After a short period of nearly constant and quantized
 288 resistance, we see a similar rise in the resistance of the quantum
 289 wire as in the case where phonon interactions were included.
 290 Since no phonon modes are available to promote or demote
 291 the electron energies, we conclude that the tunneling regime
 292 is due to large energetic spacing of longitudinal states in
 293 the channel.

294 Therefore, while there is an initial ballistic region at high
 295 drain voltages, it is interrupted by a tunneling regime. After
 296 the tunneling regime, we see a dramatic reduction to the proper
 297 ballistic quantum resistance for several propagating modes.
 298 This ballistic region then persists, in the case of this specific
 299 device, for 12.33 ± 1.41 nm, after subtracting off the length
 300 of the tunneling region, before it begins to increase linearly as
 301 expected. This is almost 10 times the length found in a similar
 302 silicon NWT at room temperature.

303 In Fig. 5, we plot the ballistic-to-diffusive crossover as a
 304 function of the drain voltage applied to the device. There
 305 is an interesting trend present in this figure. We find that
 306 the ballistic-to-diffusive crossover seems to show a nonlinear
 307 increase with increasing drain voltage. This is quite different
 308 from the linearly reducing dependence on drain voltage found
 309 in silicon. To explain how this is possible, we reexamine the
 310 forms of the POP emission and absorption terms shown in (18).
 311 For low-energy carriers, the scattering rate is appreciable over
 312 a range of grid points in the longitudinal direction. This leads
 313 to large scattering rates which significantly affects the incident
 314 carriers. However, as the energy of the carriers is increased, the
 315 phonon scattering rates become local. This localization causes
 316 a significant reduction in the strength of the electron-phonon
 317 interaction contribution to the Hamiltonian. Therefore, we ex-
 318 pect to see small ballistic-to-diffusive lengths at low drain
 319 energy **but** which increase as the carriers gain energy from the
 320 applied field. Perhaps the more unexpected part of this result is
 321 not that we obtain longer ballistic-to-diffusive lengths, but that
 322 the connection between the quasi-1-D electrons in the channel
 323 and the reservoirs produces a shift in the location of the ballistic
 324 regime. This is an important feature to note for future ultrasmall

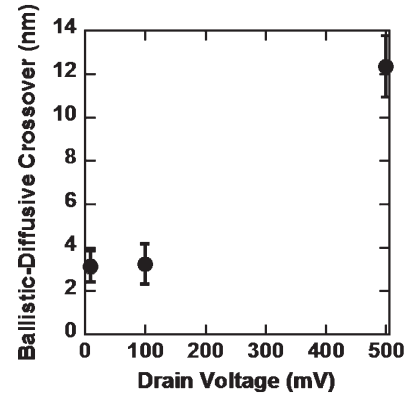


Fig. 5. Summary of the ballistic-to-diffusive crossover in InAs NWTs for different drain voltages at a constant gate voltage of $V_g = 0.5$ V. We see a nonlinear increasing trend forming in the ballistic-to-diffusive crossover length. This is due to the localization of POP scattering events.

device fabrication as it imposes serious limits on the scalability
 of NWTs.

Put in more physical terms, we should expect to find the
 ballistic length to roughly follow

$$L_{\text{ballistic}} = \frac{eVt^2}{2mL}. \quad (20)$$

Where in (20), V is the applied voltage, t is the mean scattering
 time, L is the length of the channel, and m is the effective mass.
 Hence, by increasing the drain bias from 10 to 500 mV, we
 should find a corresponding increase of the ballistic length of
 50 times. The fact that we do not see such a large increase is due
 to the fact that the mean scattering time is reduced significantly
 by hot carriers for the higher drain bias. This results in the
 value of about 12 nm for the ballistic length which is only an
 increase of 3.3 times compared to that of the low-voltage result.
 The reduction in the expected ballistic length corresponds to
 a decrease in the mean scattering time from 18 fs to about
 8.5 fs. While these times are very short, this is indicative of the
 fact that the transport is not being conducted in a bulk sample.
 Using the bulk value of the mobility of InAs, we find that the
 mean scattering time is around 180 fs (assuming a mobility of
 $14000 \text{ cm}^2/\text{V} \cdot \text{s}$). If the carriers reach the saturation velocity,
 and drop the entire bias energy into longitudinal optical (LO)
 phonons, then the estimated mean scattering time would only
 be about 3.6 fs. Thus, the carriers are dropping a great deal of
 energy in the drain rather than in the channel of the device and
 our scattering time estimates seem to be in order.

B. Temperature Dependence

In Fig. 6, we once again plot the resistance of an InAs
 quantum wire device as a function of the channel length. As
 expected, as the device is cooled, the interactions with the
 phonon modes in the device should be reduced. We see that the
 initial 12 nm of the device exhibits the same type of behavior
 as in the higher temperature cases, however, now the tunneling
 regime now takes on a more pronounced appearance. This is
 due to the fact that there is no longer enough energy to broaden
 the levels in the channel and the energetic distribution of

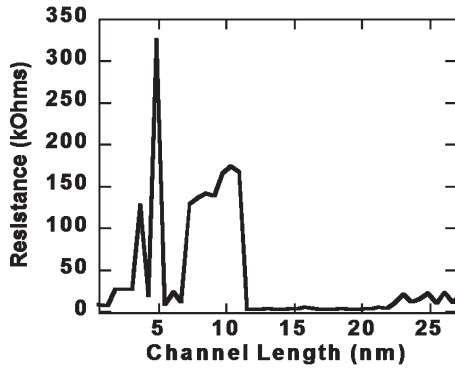


Fig. 6. Resistance of an InAs quantum wire as a function of the length of the channel. The gate voltage has been set to $V_g = 0.5$ V and the drain voltage has been set to $V_d = 0.5$ V. The temperature of the device is 100 K.

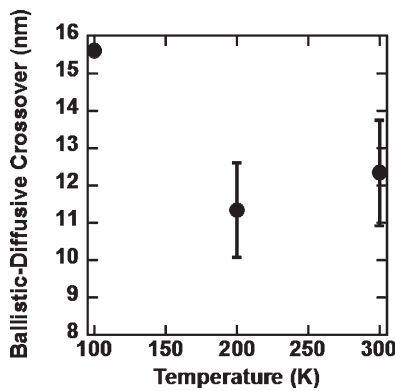


Fig. 7. Summary of the ballistic to diffusive crossover in InAs NWTs for temperatures at a constant voltages of $V_g = V_d = 0.5$ V. We see a decreasing nonlinear trend, within error bars, forming in the ballistic to diffusive crossover length. There are no error bars on the 100 K case as it is an estimate.

360 electrons becomes sharper. These effects cause the oscillations
361 in conductance to become more pronounced. At temperatures
362 this low, it is quite difficult to define a ballistic-to-diffusive
363 length as the resistance never seems to begin to exhibit linear
364 behavior. While the temperature is low, there is still enough
365 thermal energy in the system so that the phonon modes are
366 not completely frozen out, but here we find that the ballistic-
367 to-diffusive crossover must occur at channel lengths in excess
368 of 15.6 nm.

369 In Fig. 7, we summarize the performance of InAs NWT
370 at several different temperatures. We find that the trend in
371 the plot shows a quickly decaying dependence. Clearly, the
372 mobility increases as the temperature is reduced. However, even
373 at 100 K, POP scattering still dominates the transport. When
374 this is coupled with the reduced size of the InAs NWT, we
375 find that the scattering associated with POP phonons increases
376 due to increased overlap integrals. Therefore, even at reduced
377 temperatures the electron-phonon interactions are crucial to
378 understanding the performance limits of III-V NWTs. It should
379 be noted that these simulations do not account for the effect of
380 carrier heating near the drain of the device. With a relatively
381 large drain voltage of $V_d = 0.5$ V applied to the NWT, carrier
382 heating will undoubtedly play a nontrivial role in determining the
383 principles of thermal operation.

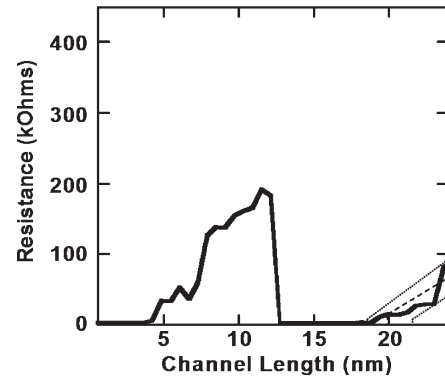


Fig. 8. Plot of the resistance of an InSb NWT as a function of the channel length with $V_g = V_d = 0.5$ V. We show the maximum and median lines used to estimate the ballistic to diffusive crossover.

C. Material Dependence

384

As mentioned in the introduction, InAs is not the only III-V
385 semiconductor currently under investigation for future use in
386 CMOS architectures; InSb and InP are also being considered.
387 In Fig. 8, we plot the resistance of an InSb NWT as a function
388 of channel length with the gate and drain voltages affixed at
389 0.5 V. The thickness of the InSb film is 9.07 nm and the width
390 of the channel is 8.42 nm. The source and drain are n -type with
391 doping set to $1.7 \times 10^{18} \text{ cm}^{-3}$ while the channel is undoped
392 p -type. While the dimensions have again been chosen to facil-
393 itate the inclusion of the discrete doping in the source and the
394 drain of the device, we have chosen very similar dimensions
395 to the InAs devices to facilitate comparison. The overall width
396 of the source and drain are 27.21 nm and the overall length
397 of the device is 50.54 nm. In the InSb device, we see very
398 similar trends as to those seen in the InAs devices. This is to be
399 expected as their POP energies are quite similar. Nevertheless,
400 while the overall shape of the curve remains similar, there are
401 some notable differences. The location of the tunneling regime
402 has shifted its position in channel length. This shift is attributed
403 to the readjustment of the states in the channel due to the
404 changes in effective mass between InAs and InSb. Therefore,
405 the initial ballistic region associated with direct tunneling is
406 estimated to be 4.73 ± 0.82 nm. After the tunneling region, we
407 find a smooth ballistic region which extends 8.7 ± 0.60 nm,
408 after subtracting off the tunneling region length. While the
409 ballistic length in InSb and InAs NWTs are quite similar,
410 the InAs device has a longer continuous region of ballistic
411 transport. Based on these results, it seems that InAs and InSb
412 are rather interchangeable.
413

Both InAs and InSb are starkly contrasted by the behavior
414 of InP. In Fig. 9, we plot the resistance of an InP NWT as a
415 function of the channel length for $V_g = V_d = 0.5$ V at 300 K.
416 The dimensions of the device are slightly different due to the
417 change in lattice constant. Here, we use a overall source and
418 drain width of 27.58 nm with a total device length of 50.47 nm.
419 The InP film thickness is 9.39 nm with a corresponding channel
420 width of 8.22 nm. The channel is assumed to be undoped p -type
421 and the source and drain are doped n -type $2 \times 10^{18} \text{ cm}^{-3}$. In
422 InP, we find that there is no significant ballistic transport regime
423 and even very short NWTs show diffusive transport. This can
424

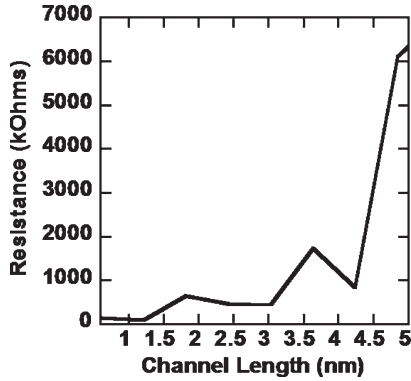


Fig. 9. Plot of the resistance of an InP NWT as a function of the channel length with $V_g = V_d = 0.5$ V. This material exhibits no ballistic regime under the conditions that we examined.

425 be explained in two ways. First, the energy of the POPs in InP
426 is much larger than in InAs or InSb. When this is combined
427 with the larger effective mass of InP, the POP scattering rate
428 is quite large and ballistic transport cannot be obtained. More
429 importantly is the fact that InP demonstrates that the channel of
430 these NWT devices are strongly coupled to the source and the
431 drain. This affects transport in that the electrons are not injected
432 directly into the channel of the device. Before they reach the
433 channel, they must first traverse the source. Therefore, by the
434 time the electrons have reached the channel of the device, they
435 have already undergone diffusive transport and this has affected
436 the mean free path in InP. Almost certainly, it is possible for InP
437 to exhibit ballistic characteristics in other situations, but not in
438 the context of the device presented here.

439 D. Effect of Discrete Doping

440 In Fig. 10(a), we plot the I_d-V_g curve for three discretely
441 doped trigate InAs quantum wire devices with only elastic
442 (boundary and impurity) scattering considered (solid) and with
443 phonon processes (dotted) while in Fig. 10(b), we repeat the
444 plots again on a logarithmic scale. For device simulation, we
445 have increased the source and drain doping to $6 \times 10^{18} \text{ cm}^{-3}$
446 and set the channel length to 9.69 nm. The device temperature
447 is 300 K and the drain voltage is held at $V_d = 0.6$ V. In terms
448 of simple device performance, averaging over four different
449 devices, we find that the threshold voltage is found to be
450 $0.373 \text{ V} \pm 11 \text{ mV}$. These figures are slightly different from
451 those quoted elsewhere [9], but it should be noted that we have
452 used a smaller sampling of devices in this paper. While the
453 threshold voltage is, at present, too high for next generation
454 technology, with the use of gate stack engineering [7] this
455 figure can be brought down to acceptable levels. Further, we
456 see that the spread in the threshold voltage is quite small. This
457 is due to the fact that fewer dopant atoms are required to reach
458 these levels of performance. Thus, with fewer elastic scattering
459 sites to induce quantum interference, the threshold voltage is
460 expected to be more stable than in its silicon-based equivalents.
461 Upon further examination of the curves in Fig. 10(b), we find
462 that in the low gate voltage regime POP scattering significantly
463 reduces the current in all three cases. This shift in low gate
464 voltage current leads to modifications to the performance. The

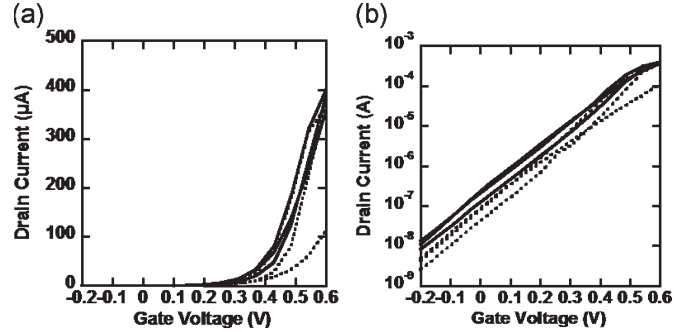


Fig. 10. (a) I_d-V_g curves for the elastic (solid) and a combination of elastic and inelastic (dotted) processes plotted on a linear scale. (b) I_d-V_g curves for the elastic (solid) and a combination of elastic and inelastic (dotted) processes plotted on a logarithmic scale.

threshold voltage averaged over four devices is $0.423 \text{ V} \pm 465$
7 mV. Based on this result, we find that the POP scattering 466
does indeed shift the threshold voltage, but it also reduces V_t 467
nonuniformities by reducing spread in carrier energy in the 468
source. At high gate voltages, we find that the drain currents 469
in both the elastic and combined (elastic scattering contribution 470
+ inelastic scattering contribution) cases begin to merge after 471
 $V_g = 0.5$ V. This can be most directly attributed to a high- 472
energy localization of the POP scattering rate which limits the 473
effect of the inelastic perturbations to current. The end result 474
is that the elastic interactions dominate the current-voltage 475
characteristics. 476

While in Fig. 10, the majority of the dopant atoms were 477
located near the bottom of the InAs layer allowing the de- 478
vice to recover quasi-ballistic behavior, one device shows a 479
very distinct difference between the ballistic and quasi-ballistic 480
cases. Here, the locations of the dopant atoms are such that in 481
the source of the device there are two dopant atoms near the 482
entrance to the channel in the middle of the semiconductor 483
layer. These dopant atoms cause significant modifications to 484
the energy of the incident electrons. The lower energy electrons 485
now see more POP scattering which, when combined with the 486
typical reflections from the channel entrance, gives rise to sig- 487
nificant reductions in the amount of charge in the channel. With 488
less conduction in the channel, the device sees significantly 489
degraded performance. 490

In Fig. 11, we plot the electron density taken at a depth of 491
approximately 5 nm into the InAs device layer at $V_g = 0.6$ V 492
for the elastic case. The black dots in the figure represent the 493
locations of the dopant atoms in the system. Dots that are 494
larger in size are closer to the surface of the device, while 495
dots that are smaller are buried farther down in the device. 496
At this voltage, we are above the threshold voltage for the 497
device; the channel is now heavily populated with carriers. 498
In the source and drain, the location of the electron density 499
depends rather weakly on the locations of the dopants. This 500
is mainly due to the scarcity of the dopants and that they are 501
buried deeply in the InAs substrate which reduces their effect 502
on the propagating electrons. Nevertheless, the density tends to 503
roughly follow the path of the dopants as it makes its way to the 504
channel. In the channel, we find that the some of the electrons 505
have been trapped by multiple sequential reflections off of the 506

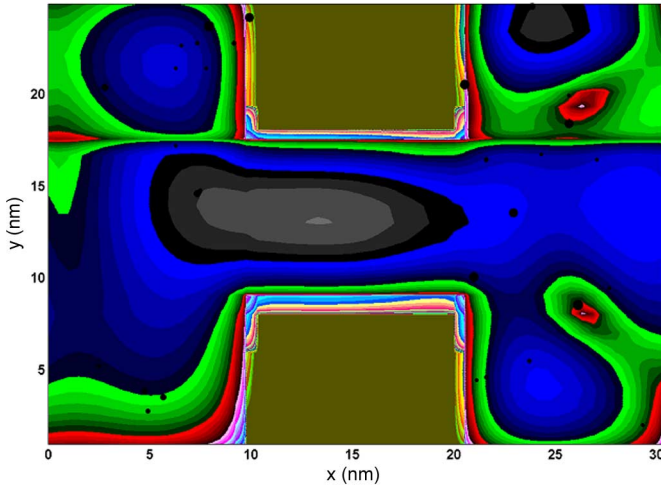


Fig. 11. Plot of the electron density in the xy plane taken at a depth of 5 nm into the InAs semiconductor layer at $V_g = V_d = 0.6$ V. The black dots represent the locations of the dopant atoms in the source and drain of the device. Larger dots represent atoms that are closer to the surface while smaller dots are more deeply buried in the semiconductor layer. At these biases, the channel is fully populated and the quantum interference is minimal.

507 source-channel and channel-drain interfaces. As the electrons
508 then exit the channel, they begin to populate sites in the drain
509 that are energetically preferential.

510 In Fig. 12, we plot the density at $V_g = 0.6$ V for the combined
511 (elastic + inelastic) case. We see very distinct differences in
512 the location and magnitude of the electrons in the combined
513 case, but, in general, it follows the same general trend as in the
514 elastic case. With the combined case, we find that there is a
515 pronounced buildup of charge in the area of the source near the
516 channel entrance of $0.38 k_b T$. The reflection off of this interface
517 alters the energy of the carriers to the point where some of
518 the lower energy modes will see increased scattering and a
519 reduction in density of approximately $0.25 k_b T$ in most areas.
520 In the source, we find additional discrepancies as this is the
521 location where the scattering is most prevalent. In the channel
522 and drain of the device, we find that there is a corresponding
523 reduction of electron density as much of the incident density
524 has been reflected at the channel interface.

525

V. CONCLUSION

526 In this paper, we have presented a method for including
527 the nonlocal effects of POP scattering into simulation tech-
528 niques relying on a propagating Hamiltonian. We then used
529 this method to simulate III-V NWTs while including the
530 effects of not only POP scattering, but also impurity, intervalley,
531 and acoustic phonon scattering. In InAs devices, we find that
532 NWTs have a significant ballistic length at room temperature
533 of about 12.33 nm. This ballistic length increases nonlinearly
534 as the drain voltage increases as the nonlocal POPs become
535 localized at higher energy. Further, we find that reductions in
536 temperature will not improve the ballistic-to-diffusive crossover
537 length, but cooling the device to around 100 K improves the
538 performance greatly due to weak phonon interactions. InSb
539 devices have similar device performance, with a ballistic-
540 to-diffusive crossover of about 8.7 nm. The great difference

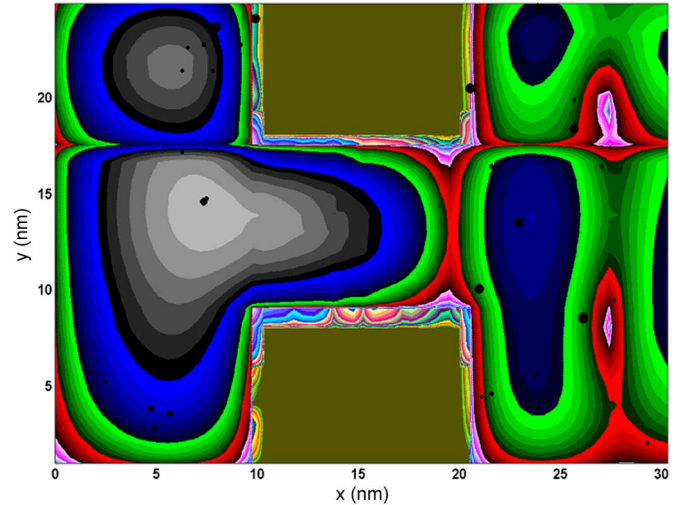


Fig. 12. Plot of the electron density in the xy plane taken at a depth of 5 nm into the InAs semiconductor layer at $V_g = V_d = 0.6$ V. The black dots represent the locations of the dopant atoms in the source and drain of the device. Here, there is significant charge build-up near the source-channel interface due to the combination of interference with the dopants near this interface and POPs.

between InAs and InSb is the locations of longitudinal states
541 in the channel which shifts a highly resistive region responsible
542 for poor device performance in both materials. This is not
543 the case with InP where we find that the POP scattering is
544 too strong and diffusive transport is achieved in the source of
545 the device. We simulated several 10-nm InAs devices to find
546 that at high gate voltages they behave as ballistic devices, but
547 the placement of dopants in the source and drain is crucial
548 to device performance. These results demonstrate the potential
549 and limitations of III-V NWTs for use in highly scaled CMOS
550 architectures. With ballistic lengths nine times longer than
551 silicon, they are certainly worthy of continued exploration for
552 novel electron quantum interference and waveguide devices. 553

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554

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REFERENCES

557

- [1] B. S. Doyle, S. Datta, M. Doczy, S. Hareland, B. Jin, J. Kavalieros, T. Linton, A. Murthy, R. Rios, and R. Chau, "High performance fully-depleted tri-gate CMOS transistors," *IEEE Electron Device Lett.*, vol. 24, no. 4, pp. 463–465, Apr. 2003. 558
- [2] T. Ashley, A. B. Dean, C. T. Elliott, G. J. Pryce, A. D. Johnson, and H. Willis, "Uncooled high-speed InSb field-effect transistors," *Appl. Phys. Lett.*, vol. 66, no. 4, pp. 481–483, Jan. 1995. 562
- [3] T. Bryllert, L.-E. Wernersson, L. E. Fröberg, and L. Samuelson, "Vertical high-mobility wrap-gated InAs nanowire transistor," *IEEE Electron Device Lett.*, vol. 27, no. 5, pp. 323–325, May 2006. 566
- [4] M. J. Gilbert and D. K. Ferry, "Efficient quantum three-dimensional modeling of fully-depleted ballistic silicon-on-insulator metal-oxide-semiconductor field-effect transistors," *J. Appl. Phys.*, vol. 95, no. 12, pp. 7954–7960, Jun. 2004. 570
- [5] —, "Quantum interference in fully-depleted tri-gate quantum wire transistors—The role of inelastic scattering," *IEEE Trans. Nanotechnol.*, vol. 4, no. 5, pp. 599–604, Sep. 2005. 573
- [6] J. Wang, E. Polizzi, and M. Lundstrom, "A three-dimensional quantum simulation of silicon nanowire transistors with the effective-mass approximation," *J. Appl. Phys.*, vol. 96, no. 4, pp. 2192–2203, Aug. 2004. 575

- 578 [7] M. J. Gilbert and D. K. Ferry, "Indium arsenide quantum wire trigate
579 metal oxide semiconductor field effect transistor," *J. Appl. Phys.*, vol. 99,
580 no. 12, p. 054503, 2006.
- 581 [8] M. J. Gilbert, R. Akis, and D. K. Ferry, "Phonon assisted ballistic to
582 diffusive crossover in silicon nanowire transistors," *J. Appl. Phys.*, vol. 98,
583 no. 9, p. 094303, Nov. 2005.
- 584 [9] M. J. Gilbert and D. K. Ferry, "Discrete dopant effects in ultra-small fully
585 depleted ballistic SOI MOSFETs," *Superlattices Microstruct.*, vol. 34,
586 no. 3–6, pp. 277–282, Sep. 2003.
- 587 [10] D. D. Johnson, "Modified Brody's method for accelerating convergence
588 in self-consistent calculations," *Phys. Rev. B, Condens. Matter*, vol. 38,
589 no. 18, pp. 12807–12813, Dec. 1988.
- 590 [11] Y. Wang, J. Wang, H. Guo, and E. Zaremba, "Many electron effects
591 on ballistic transport," *Phys. Rev. B, Condens. Matter*, vol. 52, no. 4,
592 pp. 2738–2746, Jul. 1995.
- 593 [12] V. A. Sablikov, S. V. Polyakov, and M. Büttiker, "Charging effects in a
594 quantum wire with leads," *Phys. Rev. B, Condens. Matter*, vol. 61, no. 20,
595 pp. 13763–13773, May 2000.
- 596 [13] G. Kirczenow, "Resonant conduction in ballistic quantum channels,"
597 *Phys. Rev. B, Condens. Matter*, vol. 39, no. 14, pp. 10452–10455,
598 May 1989.
- 599 [14] A. T. Tilke, F. C. Simmel, H. Lorenz, R. H. Blick, and J. P. Kotthaus,
600 "Quantum interference in a one-dimensional silicon nanowire," *Phys. Rev.*
601 *B, Condens. Matter*, vol. 68, no. 7, p. 075311, Aug. 2003.



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Ballistic to Diffusive Crossover in III–V Nanowire Transistors

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Abstract—In this paper, we examine the crossover between ballistic and diffusive transport in III–V nanowire transistors. We find that at lower drain voltages the ballistic-to-diffusive crossover occurs at channel lengths of approximately 2.3 nm at room temperature. However, when we increase the drain voltage, we find that the ballistic-to-diffusive crossover can be more than nine times as long at room temperature. As the temperature is decreased, we find that the initially the performance is not significantly improved. However, as the temperature approaches 100 K, the ballistic-to-diffusive crossover increases to longer channel lengths quite dramatically. When InAs, InSb, and InP nanowires are compared at room temperature, we find that InAs and InSb perform in a similar fashion each with ballistic regions in excess of 10 nm, but that InP has no significant ballistic regime. Finally, we simulate several 10-nm InAs trigate transistors and show that for dopants deeply buried in the source and drain the devices appear ballistic, but when dopants appear near the source-channel interface, significant reductions in performance occur.

Index Terms—Nanowire transistors (NWTs), phonon interactions, quantum simulation, III–V.

I. INTRODUCTION

THE NANOWIRE transistor (NWT) [1] may be a viable device to replace bulk Si MOSFETs due to its ability to reduce short-channel effects and give greater control over the channel electron density. There has also been a resurgence in the exploration of III–V materials for next generation CMOS technology. While silicon is the unquestioned material of choice in CMOS technology, III–V materials have some advantages over silicon such as higher mobility, which could potentially lead to very high-speed low-power transistors. Of course, we recognize the tradeoff in terms of low density-of-states, i.e., Landauer channels. InSb has measured electron mobilities of $7.8 \times 10^4 \text{ cm}^2 \text{ V}^{-1} \cdot \text{s}^{-1}$ [2] and has long been touted as a good candidate for high-frequency devices. Recently, high-mobility wrap-gated p-channel InAs NWTs have been experimentally demonstrated which also have exceptionally high mobilities and solid device performance [3]. With the advent of technology capable of producing high quality III–V nanowires for transistor applications, it is natural to question the limits of device performance in III–V materials.

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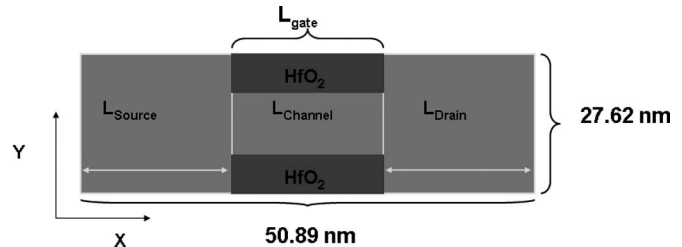


Fig. 1. Schematic of the device used in this paper taken in the xy plane.

Of particular interest for future generation CMOS applications is the location of the ballistic-to-diffusive crossover in III–V nanowire systems.

There have been many previous studies demonstrating the performance of silicon NWT [4]–[6] and InAs NWT [7]. However, since we are interested in operating these devices at or near room temperature, phonon processes are very important. A recent study has placed the ballistic-to-diffusive crossover for silicon nanowire systems at 1.42 nm [8]. This places serious limits on the effectiveness of silicon in next generation CMOS but it also shows the importance of phonon modes in the determination of the performance characteristics. In this paper, we examine the dependence of the ballistic-to-diffusive (mobility dominated transport) crossover point in III–V NWT as a function of temperature, drain voltage, and semiconductor material covering three of the most popular III–V materials: InAs, InSb, and InP using a fully 3-D self-consistent quantum transport formalism. The effects of phonon scattering are included in the simulations as separable self-energy terms in the Hamiltonian. To make the simulations as complete as possible, we include interactions with the dopant ions, and inelastic effects (the effects of acoustic deformation potential, intervalley phonons [$\Gamma \rightarrow X$ and $\Gamma \rightarrow L$], and polar optical phonons (POPs)].

II. DEVICE STRUCTURE AND SIMULATION METHOD

In Fig. 1, we display a schematic of the system under consideration in the x – y plane (the z -axis is normal to the plane shown). The device dimensions (in multiples of the lattice constant) use in the simulations were chosen to aid in the inclusion of the discrete dopants. The thickness of the InAs layer is 9.09 nm and the width of the channel is 8.48 nm. The source and drain of the device are n -type with a doping density of $1.3 \times 10^{18} \text{ cm}^{-3}$, while the channel of the device is considered to be p -type, but undoped. The gate material is assumed to be platinum and the gate oxide on each side is 1 nm of hafnium oxide (HfO_2). Underneath the device, we

79 have assumed a thick silicon dioxide substrate. The length of
80 the device is set to 50.89 nm. While the channel of the device
81 is typically much smaller than the overall length of the device
82 geometry, a large domain is set so that as the length of the
83 channel is increased, the bandstructure for the overall device
84 remains constant. The length of the channel starts at a length
85 of 6.0583 Å and is increased by this length until the diffusive
86 regime is obtained. During this increase of the length, the width,
87 and depth of the nanowire channel are kept constant. The length
88 of the gate that surrounds the nanowire on three sides is set to
89 be the same length as the channel of the device. The source and
90 drain of the device are divided equally depending on the length
91 of the channel; however, they never are reduced below 10 nm
92 in the x -direction.

93 Once the device geometry is defined, the InAs lattice is
94 scanned and the dopants are distributed according to the method
95 presented in [9]. Following the distribution of the dopants, they
96 are then mapped back onto the grid of the simulation mesh and
97 the initial self-consistent Poisson solution is obtained. In this
98 case, the full Coulomb potential of the dopants is incorporated.
99 Then, the solution of Poisson's equation for the local potential
100 is no longer smoothly varying in the source and drain of the
101 device. The inclusion of discrete dopants causes the formation
102 of potential variations in the source and drain. The density
103 throughout the device is calculated using a variant of the re-
104 cursive scattering matrix method which solves the Schrödinger
105 equation in the effective mass approximation, as described in
106 [4]. In short, this method provides us with a 3-D, fully quantum
107 mechanical treatment. Since the transport calculation is per-
108 formed in real space, the different excited modes in the system
109 are automatically coupled unlike in other simulations [6]. In
110 order to achieve self-consistency, the density obtained from the
111 transport calculation is then updated using Broyden's method
112 [10] and a new guess for the potential is obtained through
113 the solution of Poisson's equation. The process is repeated
114 until a desired level of convergence is obtained. Exchange and
115 correlations terms are included in each simulation through a
116 local density approximation [11].

117 In this paper, we seek the ballistic-to-diffusive crossover in
118 these nanowire systems. Our procedure for determining this
119 quantity is given below. We know that even when the transport
120 is ballistic, the lower limit of the resistance of the channel is
121 determined by the inverse of the Landauer conductance

$$R_{\text{ballistic}} = \left[\frac{2e^2}{h} N \right]^{-1} \quad (1)$$

122 where N is the number of transverse modes flowing in the
123 quantum wire. Clearly, the resistance in (1) has no dependence
124 on the length of the quantum wire. Conversely, when the
125 transport is determined by the carrier mobility and the carrier
126 density, then the resistance becomes

$$R_{\text{diffusive}} = \frac{1}{ne\mu} \frac{L}{A} \quad (2)$$

127 where L is the channel length and A is the cross-sectional
128 area of the inversion layer. We have been careful to use the

area of the inversion layer rather than the width because we
129 are dealing with a 3-D quantum wire with quantization in the
130 transverse directions. In this regime, there is a clear linear de-
131 pendence of the resistance with the length of the quantum wire
132 channel. 133

III. POP SCATTERING

134

The inclusion of separable scattering mechanisms is not a
135 new concept. In previous work, acoustic and optical phonon
136 processes have been derived [8] and so the work will not be
137 repeated here. However, we are now interested in including the
138 nonlocal effects of POPs which, to this point, have not been
139 treated. This is a crucial difference between silicon and III-V
140 materials. In silicon, the local phonon processes dominate,
141 but in III-V materials nonlocal optical processes dominate the
142 room temperature transport. We begin by noting that to use the
143 method presented in [8], we are assuming that the scattering
144 is weak relative to the energies in the system, and that we
145 are only interested in the steady-state limit. Therefore, we can
146 use the Fermi golden rule expression, equivalent to a first-
147 order nonself-consistent Born approximation, for each of the
148 scattering processes and generate a real space self-energy from
149 it. In particular, we note that the imaginary part of the self-
150 energy term is related to the scattering rate and it is the latter
151 scattering rate that we wish to calculate [8]. This will result in
152 an x -directed momentum which is related to the carrier energy
153 in the quantum wire. While the local scattering rates have been
154 previously derived [8], we will present the important changes
155 in the general derivation approach to take into account the
156 nonlocal nature of POP scattering. In this paper, we use the
157 Fermi golden rule to calculate scattering rates. This is treated in
158 many other places, however, we must account for the transverse
159 modes of the quantum wire. Therefore, we begin with the
160 general form 161

$$\begin{aligned} \left(\frac{1}{\tau_{\text{polar}}} \right)_{ij}^{i'j'} &= \frac{\pi e^2}{\gamma V \omega_0} \sum_{q_x} \sum_{q_y, q_z} \int dy \int dz \int dy' \int dz' \\ &\times \varphi_{i,j}^*(y, z) \varphi_{i',j'}(y, z) \varphi_{i,j}^*(y', z') \varphi_{i',j'}(y', z') \\ &\times \frac{e^{iq_y(y-y') + iq_z(z-z')}}{q^2} \\ &\times \{ (N_q + 1) \delta(E_k - E_{k-q} - \hbar\omega_0) \\ &\quad + N_q \delta(E_k - E_{k-q} + \hbar\omega_0) \} \quad (3) \end{aligned}$$

where V is the volume, ω_0 is the phonon frequency and γ is
162 the effective interaction parameter, i is the initial mode index
163 in the width direction, j is the initial mode index in the depth
164 direction, i' is the final mode index in the width direction,
165 and j' is the final mode index in depth direction. The delta
166 functions in (3) serve to conserve the energy in the process
167 of the interactions of the carriers with the POPs. $\varphi_{i,j}(y, z)$ is
168 the transverse 2-D wave function in the particular slice under
169 consideration. We assume in this treatment that we are dealing
170 with parabolic bandstructure. Taking a closer look at the terms 171

172 contained within the delta functions, we realize that based on
173 this assumption

$$\begin{aligned} E_k - E_{k\pm q} \pm \hbar\omega_0 &= E_{i,j} - E_{i',j'} \pm \hbar\omega_0 + \frac{\hbar^2 k_x^2}{2m_x^*} - \frac{\hbar^2 (k_x \pm q_x)^2}{2m_x^*} \\ &= E_{i,j} - E_{i',j'} \pm \hbar\omega_0 - \frac{\hbar^2 k_x q_x}{m_x^*} \pm \frac{\hbar^2 q_x^2}{2m_x^*}. \end{aligned} \quad (4)$$

174 In (4), $E_{i,j}$ and $E_{i',j'}$ are the initial and final energies at the
175 bottom of their respective bands. We begin to simplify (3), by
176 first examining the longitudinal terms

$$\sum_{q_x} \frac{1}{2q_x^2} \delta(E_k - E_{k\pm q} \pm \hbar\omega_0). \quad (5)$$

177 We now expand (5) using the results from (4)

$$\sum_{q_x} \frac{1}{q_x^2 + q_y^2 + q_z^2} \delta\left(\mp \frac{\hbar^2 q_x^2}{2m_x^*} \mp \frac{\hbar^2 k_x q_x}{m_x^*} + E_{i,j} - E_{i',j'} \pm \hbar\omega_0\right). \quad (6)$$

178 This sum may be simplified further by taking the Fourier
179 transform with respect to q_x

$$\frac{1}{2\pi} \int_{-\infty}^{\infty} \frac{dq_x e^{iq_x(x-x')}}{q_x^2 + q_y^2 + q_z^2} \delta\left(\mp \frac{\hbar^2 q_x^2}{2m_x^*} \mp \frac{\hbar^2 k_x q_x}{m_x^*} + \Delta_{i',j'}\right) \quad (7)$$

180 where we denote

$$\Delta_{i',j'} = E_{i,j} - E_{i',j'} \pm \hbar\omega_0. \quad (8)$$

181 Representing (7) as a summation, we have

$$\begin{aligned} &\frac{1}{2\pi} \int_{-\infty}^{\infty} \frac{dq_x e^{iq_x(x-x')}}{q_x^2 + q_y^2 + q_z^2} \frac{1}{|\partial E_{k\pm q}/\partial q_x|} \\ &\frac{m_x^*}{2\pi\hbar^2} \int_{-\infty}^{\infty} \frac{dq_x e^{iq_x(x-x')}}{q_x^2 + q_y^2 + q_z^2} \frac{1}{\left|\mp k_x + \frac{q_x}{2}\right|}. \end{aligned} \quad (9)$$

182 At this point, we examine the relationship for $E_{k\pm q}$

$$E_{k\pm q} = \frac{\hbar^2 k_x^2}{2m_x^*} \pm \frac{\hbar^2 k_x q_x}{m_x^*} + \frac{\hbar^2 q_x^2}{2m_x^*}. \quad (10)$$

183 Equation (10) is quadratic in q_x and may then be solved to yield
184 a solution for q_x which is then substituted back into (9) making
185 the term in the absolute value independent of q_x as shown
186 below regardless of if the phonon process under consideration
187 is emission or absorption. Therefore, we may group all of
188 the variables that are independent of q_x into a simple dummy
189 variable a . With this pairing, we may reduce our summation in
190 (9) to reach

$$\frac{m_x^*}{2\pi\hbar^2} \int_{-\infty}^{\infty} \frac{dq_x e^{iq_x(x-x')}}{q_x^2 + a^2}. \quad (11)$$

Equation (11) is most easily evaluated by making a transforma-
tion to cylindrical coordinates to obtain

$$\frac{m_x^*}{4\pi\hbar^2} \iint \frac{q_x e^{iq_x(\vec{r} - \vec{r}')}}{q_{xi}^2 + a^2} d\theta dq. \quad (12)$$

We evaluate the integral in (12) using contour integrations to
arrive at a final, simplified expression for the sum over the
longitudinal wavevectors

$$\sum_{q_x} \frac{1}{2q_x} \delta(E_k). \quad (13)$$

The expression in (13) is substituted back into (3) to find an
overlap integral that is similar to both the one found in [8] and
in most textbooks

$$\begin{aligned} &= \sum_{q_x} \sum_{q_y, q_z} \int dy \int dz \int dy' \int dz' \varphi_{i,j}^*(y, z) \varphi_{i',j'}(y, z) \\ &\quad \times \varphi_{i,j}^*(y', z') \varphi_{i',j'}(y', z') \frac{e^{iq_y(y-y')} + iq_z(z-z')}{2q_x} \delta(E_k) \\ &= \frac{A}{2V} \sum_{q_x} \iint dz dy [\varphi_{i,j}^*(y, z) \varphi_{i',j'}(y, z)]^2 \frac{1}{q_x} \delta(E_k) \\ &= \sum_{k_x} I_{i,j}^{i',j'} \frac{1}{k_x} \delta(E_k). \end{aligned} \quad (14)$$

Where in (14), we have used the following definition

$$I_{i,j}^{i',j'} = \frac{A}{4\pi V} \iint dy dz [\varphi_{i,j}^*(y, z) \varphi_{i',j'}(y, z)]^2. \quad (15)$$

Following the usual procedure, we replace the summation over
the final momentum states with an integration

$$\sum_{k'} \rightarrow \frac{L}{2\pi} \int_{-\infty}^{\infty} dk' = \int_0^{\infty} \rho_{1D}(E') dE' \quad (16)$$

to obtain the final result

$$\left(\frac{1}{\tau_{\text{polar}}}\right)_{i,j}^{i',j'} = \frac{m_x^* L e^2}{32\pi^3 \gamma \omega_0} I_{i,j}^{i',j'} \int \frac{dk_x}{k_x} \delta(E_k). \quad (17)$$

However, we are still in the mode space representation and
the longitudinal momentum is not yet a viable operator. From
many-body physics, the momentum dependence arises from the
Fourier transform of the differences in the two coordinates in
the self-energy. Therefore, to include the proper momentum
dependence, we take the inverse transform. This essentially
involves solving a contour integral for both the emission and
absorption cases which results from (17). This gives us our
final, simplified and summarized results for absorption and
emission of POPs

$$\left(\frac{1}{\tau_{\text{polar}}}\right)_{i,j}^{i',j'} = \frac{m_x^* L e^2}{32\pi^3 \gamma \omega_0} I_{i,j}^{i',j'} \left\{ \begin{aligned} &N_q e^{-k_0|x-x'|} \\ &(N_q + 1) \frac{\sin(k_0|x-x'|)}{k_0|x-x'|} \end{aligned} \right\}. \quad (18)$$

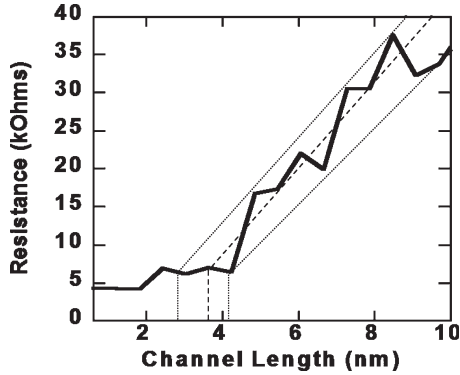


Fig. 2. Resistance of an InAs quantum wire as a function of the length of the channel. The gate voltage has been set to $V_g = 0.5$ V and the drain voltage has been set to $V_d = 10$ mV. The temperature of the device is 300 K. The red lines signify the maximum and minimum ranges for the ballistic to diffusive crossover, and the blue line signifies the median value. These lines are used to determine the value quoted for the ballistic to diffusive crossover.

213 This result is reasonable as we now see the manifestation of
 214 the nonlocal nature of POP scattering as a clear dependence on
 215 distance in (18). To utilize this in the transport calculation, we
 216 must use the following unitary transformation to convert this
 217 form to the site representation for inclusion in the Hamiltonian

$$\Gamma_{\text{polar}} = \text{Im}\{\Sigma\} = U^+ \left(\frac{\hbar}{\tau_{\text{polar}}} \right)_{i,j}^{i',j'} U \quad (19)$$

218 where U is a unitary mode-to-site transformation matrix. The
 219 unitary matrix U^+ results from the eigenvalue solutions in the
 220 transverse slice and are composed of the various eigenfunc-
 221 tions in the site basis. Hence, it represents a mode-to-slice
 222 transformation.

223 IV. RESULTS

224 A. Drain-Voltage Dependence

225 In Fig. 2, we plot the resistance of an InAs quantum wire
 226 as a function of the length of the channel for a small drain
 227 bias of $V_d = 10$ mV. In each of the following plots, the gate
 228 voltage is held constant at $V_g = 0.5$ V in order to ensure that
 229 there is a channel in the p -type wire. It should be noted that the
 230 optical phonon energy used in the calculation of InAs devices
 231 is 30 meV. While the channel length remains below 3.6 nm, the
 232 resistance of the device does not appear to change significantly.
 233 The oscillations in the resistance in Fig. 2 arise mainly from
 234 the interaction of the incident electron modes with the material
 235 boundaries, and in particular, with the source and channel
 236 interfaces creating strong quantum interference which is seen
 237 not only in theoretical studies [12], [13] but also experimentally
 238 [14]. For channels longer than about 3.6 nm, the resistance
 239 shows a clear linear dependence as the channel length increases
 240 as is expected when the nature of the transport changes from
 241 ballistic to diffusive. At this drain voltage, we estimate the
 242 ballistic-to-diffusive crossover to occur at 3.12 ± 0.3 nm. This
 243 is limited by the strength of the POP scattering as the X and
 244 L valleys are offset by sufficient energy to significantly limit

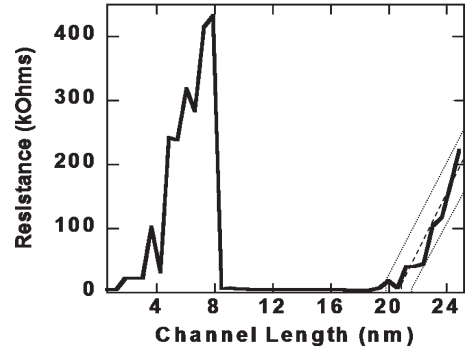


Fig. 3. Resistance of an InAs gated quantum wire as a function of the length of the channel. The gate voltage has been set to $V_g = 0.5$ V and the drain voltage has been set to $V_d = 0.5$ V. The temperature of the device is 300 K. The red lines once again signify the maximum and minimum ranges for the ballistic to diffusive crossover, and the blue line signifies the median value.

the intervalley scattering rates to these valleys. While this value
 245 is more than that of the quoted value of 1.42 nm for similar
 246 cross-sectional silicon quantum wires, this is not the significant
 247 increase we were hoping to see. Further, there will not be
 248 significant device-to-device variation in InAs or other III-V
 249 NWTs due to low concentration of dopant atoms as compared
 250 to that of a typical silicon device. 251

In Fig. 3, we now increase the drain voltage to $V_d = 0.5$ V.
 252 We find that this drain voltage consists of much more structure
 253 than the device held at reduced drain voltages. We begin our
 254 examination of Fig. 3, by noticing that there is an initial period
 255 of slightly more than 3 nm where the resistance of the quantum
 256 wire channel does not change. In this region, the length of the
 257 channel is such that the incident electrons see almost no channel
 258 potential and simply tunnel from the source to the drain of
 259 the device. As the length of the channel is increased beyond
 260 4 nm, we see that there is a large spike in the resistance of the
 261 channel. This behavior is attributed to the fact that the locations
 262 of the longitudinal states in the channel are too high in energy
 263 to support any modes and the channel becomes a large tunnel
 264 barrier. This argument is supported by the fact that the form of
 265 the drain voltage versus channel length in the regime of 4 to
 266 8 nm has an exponential form which is indicative of tunneling
 267 transport. We do not see this in the lower drain voltage sweeps
 268 due to the fact that scattering has blurred energy spectrum of
 269 the electrons before they reach the channel. This widening of
 270 the range of energies allows electrons not energetically capable
 271 of accessing the states to scatter into them. Thus, we see no rise
 272 in resistance in Fig. 2. While this type of interaction is fairly
 273 uncommon in semiconductor devices due to their relatively
 274 large sizes, this is common in systems of quasi-1-D electrons
 275 connected to external reservoirs [13]. 276

The existence of the tunneling regime in Fig. 3 is an im-
 277 portant point. If indeed, the physics is as we state then the
 278 same tunneling regime should show up in simple ballistic
 279 simulations of the same device structure. In Fig. 4, we plot
 280 the resistance of the quantum wire channel as a function of its
 281 length with no scattering included in the simulation. We find a
 282 similar trend as in the case with the inelastic processes included.
 283 In this simulation, we initially begin with three propagating
 284 modes in the system. The initial resistances do not match the 285

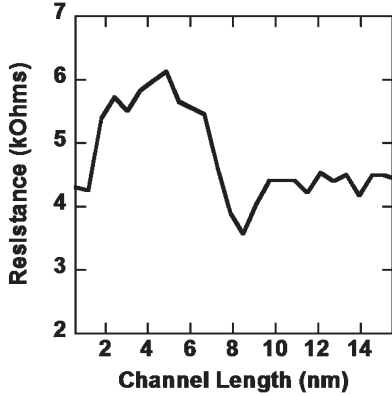


Fig. 4. Plot of the resistance of the quantum wire as a function of its length with no scattering included in the simulation for $V_g = V_d = 0.5$ V. We find the same short channel resistance trend in the ballistic case as we find in the case with the phonon processes included. Since the resistance increase in the quantum wires is the focus, we plot only the first 8 nm of channel length.

286 other quantum wires as doping is slightly lower than in other
 287 devices. After a short period of nearly constant and quantized
 288 resistance, we see a similar rise in the resistance of the quantum
 289 wire as in the case where phonon interactions were included.
 290 Since no phonon modes are available to promote or demote
 291 the electron energies, we conclude that the tunneling regime
 292 is due to large energetic spacing of longitudinal states in
 293 the channel.

294 Therefore, while there is an initial ballistic region at high
 295 drain voltages, it is interrupted by a tunneling regime. After
 296 the tunneling regime, we see a dramatic reduction to the proper
 297 ballistic quantum resistance for several propagating modes.
 298 This ballistic region then persists, in the case of this specific
 299 device, for 12.33 ± 1.41 nm, after subtracting off the length
 300 of the tunneling region, before it begins to increase linearly as
 301 expected. This is almost 10 times the length found in a similar
 302 silicon NWT at room temperature.

303 In Fig. 5, we plot the ballistic-to-diffusive crossover as a
 304 function of the drain voltage applied to the device. There
 305 is an interesting trend present in this figure. We find that
 306 the ballistic-to-diffusive crossover seems to show a nonlinear
 307 increase with increasing drain voltage. This is quite different
 308 from the linearly reducing dependence on drain voltage found
 309 in silicon. To explain how this is possible, we reexamine the
 310 forms of the POP emission and absorption terms shown in (18).
 311 For low-energy carriers, the scattering rate is appreciable over
 312 a range of grid points in the longitudinal direction. This leads
 313 to large scattering rates which significantly affects the incident
 314 carriers. However, as the energy of the carriers is increased, the
 315 phonon scattering rates become local. This localization causes
 316 a significant reduction in the strength of the electron-phonon
 317 interaction contribution to the Hamiltonian. Therefore, we ex-
 318 pect to see small ballistic-to-diffusive lengths at low drain
 319 energy **but** which increase as the carriers gain energy from the
 320 applied field. Perhaps the more unexpected part of this result is
 321 not that we obtain longer ballistic-to-diffusive lengths, but that
 322 the connection between the quasi-1-D electrons in the channel
 323 and the reservoirs produces a shift in the location of the ballistic
 324 regime. This is an important feature to note for future ultrasmall

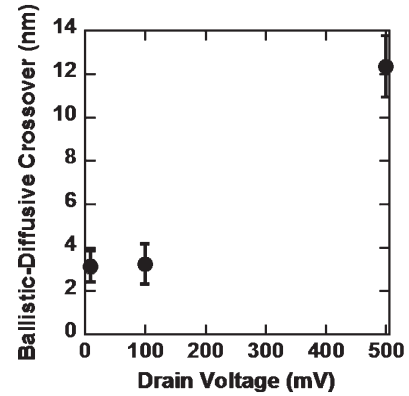


Fig. 5. Summary of the ballistic-to-diffusive crossover in InAs NWTs for different drain voltages at a constant gate voltage of $V_g = 0.5$ V. We see a nonlinear increasing trend forming in the ballistic-to-diffusive crossover length. This is due to the localization of POP scattering events.

device fabrication as it imposes serious limits on the scalability
 of NWTs.

Put in more physical terms, we should expect to find the
 ballistic length to roughly follow

$$L_{\text{ballistic}} = \frac{eVt^2}{2mL}. \quad (20)$$

Where in (20), V is the applied voltage, t is the mean scattering
 time, L is the length of the channel, and m is the effective mass.
 Hence, by increasing the drain bias from 10 to 500 mV, we
 should find a corresponding increase of the ballistic length of
 50 times. The fact that we do not see such a large increase is due
 to the fact that the mean scattering time is reduced significantly
 by hot carriers for the higher drain bias. This results in the
 value of about 12 nm for the ballistic length which is only an
 increase of 3.3 times compared to that of the low-voltage result.
 The reduction in the expected ballistic length corresponds to
 a decrease in the mean scattering time from 18 fs to about
 8.5 fs. While these times are very short, this is indicative of the
 fact that the transport is not being conducted in a bulk sample.
 Using the bulk value of the mobility of InAs, we find that the
 mean scattering time is around 180 fs (assuming a mobility of
 $14000 \text{ cm}^2/\text{V} \cdot \text{s}$). If the carriers reach the saturation velocity,
 and drop the entire bias energy into longitudinal optical (LO)
 phonons, then the estimated mean scattering time would only
 be about 3.6 fs. Thus, the carriers are dropping a great deal of
 energy in the drain rather than in the channel of the device and
 our scattering time estimates seem to be in order.

B. Temperature Dependence

In Fig. 6, we once again plot the resistance of an InAs
 quantum wire device as a function of the channel length. As
 expected, as the device is cooled, the interactions with the
 phonon modes in the device should be reduced. We see that the
 initial 12 nm of the device exhibits the same type of behavior
 as in the higher temperature cases, however, now the tunneling
 regime now takes on a more pronounced appearance. This is
 due to the fact that there is no longer enough energy to broaden
 the levels in the channel and the energetic distribution of

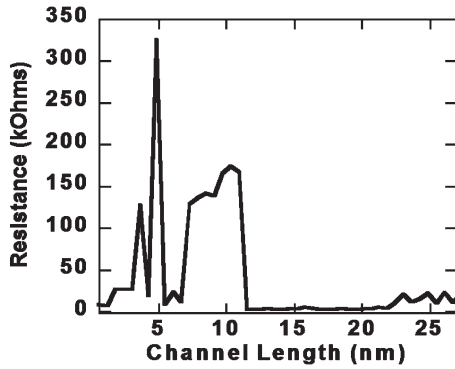


Fig. 6. Resistance of an InAs quantum wire as a function of the length of the channel. The gate voltage has been set to $V_g = 0.5$ V and the drain voltage has been set to $V_d = 0.5$ V. The temperature of the device is 100 K.

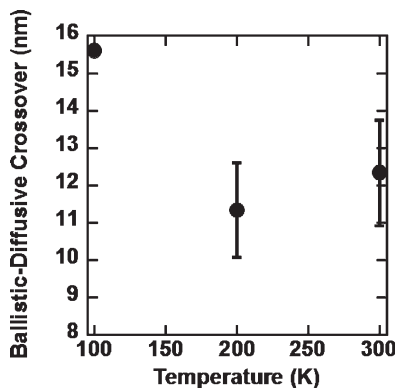


Fig. 7. Summary of the ballistic to diffusive crossover in InAs NWTs for temperatures at a constant voltages of $V_g = V_d = 0.5$ V. We see a decreasing nonlinear trend, within error bars, forming in the ballistic to diffusive crossover length. There are no error bars on the 100 K case as it is an estimate.

360 electrons becomes sharper. These effects cause the oscillations
361 in conductance to become more pronounced. At temperatures
362 this low, it is quite difficult to define a ballistic-to-diffusive
363 length as the resistance never seems to begin to exhibit linear
364 behavior. While the temperature is low, there is still enough
365 thermal energy in the system so that the phonon modes are
366 not completely frozen out, but here we find that the ballistic-
367 to-diffusive crossover must occur at channel lengths in excess
368 of 15.6 nm.

369 In Fig. 7, we summarize the performance of InAs NWT
370 at several different temperatures. We find that the trend in
371 the plot shows a quickly decaying dependence. Clearly, the
372 mobility increases as the temperature is reduced. However, even
373 at 100 K, POP scattering still dominates the transport. When
374 this is coupled with the reduced size of the InAs NWT, we
375 find that the scattering associated with POP phonons increases
376 due to increased overlap integrals. Therefore, even at reduced
377 temperatures the electron-phonon interactions are crucial to
378 understanding the performance limits of III-V NWTs. It should
379 be noted that these simulations do not account for the effect of
380 carrier heating near the drain of the device. With a relatively
381 large drain voltage of $V_d = 0.5$ V applied to the NWT, carrier
382 heating will undoubtedly play a nontrivial role in determining the
383 principles of thermal operation.

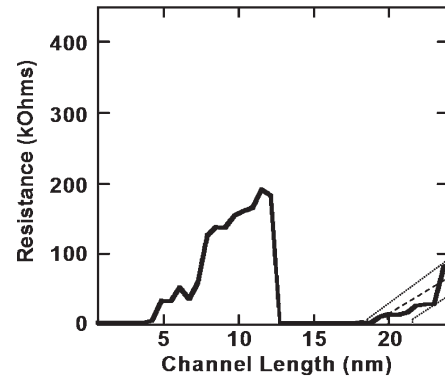


Fig. 8. Plot of the resistance of an InSb NWT as a function of the channel length with $V_g = V_d = 0.5$ V. We show the maximum and median lines used to estimate the ballistic to diffusive crossover.

C. Material Dependence

384

As mentioned in the introduction, InAs is not the only III-V
385 semiconductor currently under investigation for future use in
386 CMOS architectures; InSb and InP are also being considered.
387 In Fig. 8, we plot the resistance of an InSb NWT as a function
388 of channel length with the gate and drain voltages affixed at
389 0.5 V. The thickness of the InSb film is 9.07 nm and the width
390 of the channel is 8.42 nm. The source and drain are n -type with
391 doping set to $1.7 \times 10^{18} \text{ cm}^{-3}$ while the channel is undoped
392 p -type. While the dimensions have again been chosen to facil-
393 itate the inclusion of the discrete doping in the source and the
394 drain of the device, we have chosen very similar dimensions
395 to the InAs devices to facilitate comparison. The overall width
396 of the source and drain are 27.21 nm and the overall length
397 of the device is 50.54 nm. In the InSb device, we see very
398 similar trends as to those seen in the InAs devices. This is to be
399 expected as their POP energies are quite similar. Nevertheless,
400 while the overall shape of the curve remains similar, there are
401 some notable differences. The location of the tunneling regime
402 has shifted its position in channel length. This shift is attributed
403 to the readjustment of the states in the channel due to the
404 changes in effective mass between InAs and InSb. Therefore,
405 the initial ballistic region associated with direct tunneling is
406 estimated to be 4.73 ± 0.82 nm. After the tunneling region, we
407 find a smooth ballistic region which extends 8.7 ± 0.60 nm,
408 after subtracting off the tunneling region length. While the
409 ballistic length in InSb and InAs NWTs are quite similar,
410 the InAs device has a longer continuous region of ballistic
411 transport. Based on these results, it seems that InAs and InSb
412 are rather interchangeable.
413

Both InAs and InSb are starkly contrasted by the behavior
414 of InP. In Fig. 9, we plot the resistance of an InP NWT as a
415 function of the channel length for $V_g = V_d = 0.5$ V at 300 K.
416 The dimensions of the device are slightly different due to the
417 change in lattice constant. Here, we use a overall source and
418 drain width of 27.58 nm with a total device length of 50.47 nm.
419 The InP film thickness is 9.39 nm with a corresponding channel
420 width of 8.22 nm. The channel is assumed to be undoped p -type
421 and the source and drain are doped n -type $2 \times 10^{18} \text{ cm}^{-3}$. In
422 InP, we find that there is no significant ballistic transport regime
423 and even very short NWTs show diffusive transport. This can
424

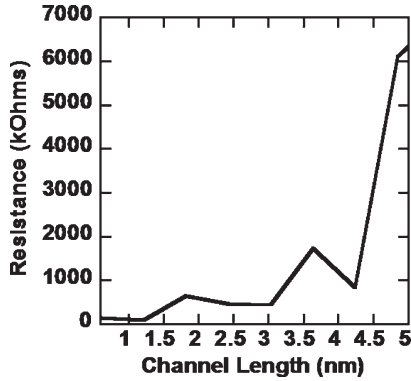


Fig. 9. Plot of the resistance of an InP NWT as a function of the channel length with $V_g = V_d = 0.5$ V. This material exhibits no ballistic regime under the conditions that we examined.

425 be explained in two ways. First, the energy of the POPs in InP
426 is much larger than in InAs or InSb. When this is combined
427 with the larger effective mass of InP, the POP scattering rate
428 is quite large and ballistic transport cannot be obtained. More
429 importantly is the fact that InP demonstrates that the channel of
430 these NWT devices are strongly coupled to the source and the
431 drain. This affects transport in that the electrons are not injected
432 directly into the channel of the device. Before they reach the
433 channel, they must first traverse the source. Therefore, by the
434 time the electrons have reached the channel of the device, they
435 have already undergone diffusive transport and this has affected
436 the mean free path in InP. Almost certainly, it is possible for InP
437 to exhibit ballistic characteristics in other situations, but not in
438 the context of the device presented here.

439 D. Effect of Discrete Doping

440 In Fig. 10(a), we plot the I_d-V_g curve for three discretely
441 doped trigate InAs quantum wire devices with only elastic
442 (boundary and impurity) scattering considered (solid) and with
443 phonon processes (dotted) while in Fig. 10(b), we repeat the
444 plots again on a logarithmic scale. For device simulation, we
445 have increased the source and drain doping to $6 \times 10^{18} \text{ cm}^{-3}$
446 and set the channel length to 9.69 nm. The device temperature
447 is 300 K and the drain voltage is held at $V_d = 0.6$ V. In terms
448 of simple device performance, averaging over four different
449 devices, we find that the threshold voltage is found to be
450 $0.373 \text{ V} \pm 11 \text{ mV}$. These figures are slightly different from
451 those quoted elsewhere [9], but it should be noted that we have
452 used a smaller sampling of devices in this paper. While the
453 threshold voltage is, at present, too high for next generation
454 technology, with the use of gate stack engineering [7] this
455 figure can be brought down to acceptable levels. Further, we
456 see that the spread in the threshold voltage is quite small. This
457 is due to the fact that fewer dopant atoms are required to reach
458 these levels of performance. Thus, with fewer elastic scattering
459 sites to induce quantum interference, the threshold voltage is
460 expected to be more stable than in its silicon-based equivalents.
461 Upon further examination of the curves in Fig. 10(b), we find
462 that in the low gate voltage regime POP scattering significantly
463 reduces the current in all three cases. This shift in low gate
464 voltage current leads to modifications to the performance. The

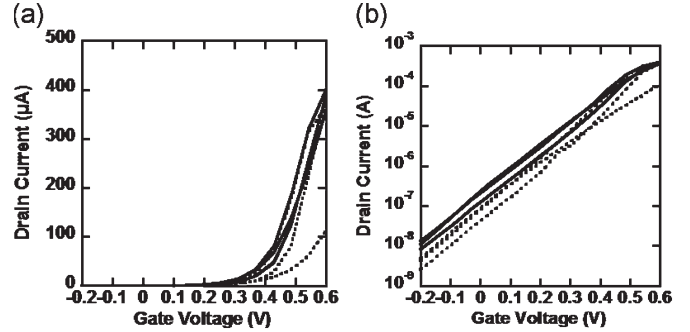


Fig. 10. (a) I_d-V_g curves for the elastic (solid) and a combination of elastic and inelastic (dotted) processes plotted on a linear scale. (b) I_d-V_g curves for the elastic (solid) and a combination of elastic and inelastic (dotted) processes plotted on a logarithmic scale.

threshold voltage averaged over four devices is $0.423 \text{ V} \pm 465$
7 mV. Based on this result, we find that the POP scattering 466
does indeed shift the threshold voltage, but it also reduces V_t 467
nonuniformities by reducing spread in carrier energy in the 468
source. At high gate voltages, we find that the drain currents 469
in both the elastic and combined (elastic scattering contribution 470
+ inelastic scattering contribution) cases begin to merge after 471
 $V_g = 0.5$ V. This can be most directly attributed to a high- 472
energy localization of the POP scattering rate which limits the 473
effect of the inelastic perturbations to current. The end result 474
is that the elastic interactions dominate the current-voltage 475
characteristics. 476

While in Fig. 10, the majority of the dopant atoms were 477
located near the bottom of the InAs layer allowing the de- 478
vice to recover quasi-ballistic behavior, one device shows a 479
very distinct difference between the ballistic and quasi-ballistic 480
cases. Here, the locations of the dopant atoms are such that in 481
the source of the device there are two dopant atoms near the 482
entrance to the channel in the middle of the semiconductor 483
layer. These dopant atoms cause significant modifications to 484
the energy of the incident electrons. The lower energy electrons 485
now see more POP scattering which, when combined with the 486
typical reflections from the channel entrance, gives rise to sig- 487
nificant reductions in the amount of charge in the channel. With 488
less conduction in the channel, the device sees significantly 489
degraded performance. 490

In Fig. 11, we plot the electron density taken at a depth of 491
approximately 5 nm into the InAs device layer at $V_g = 0.6$ V 492
for the elastic case. The black dots in the figure represent the 493
locations of the dopant atoms in the system. Dots that are 494
larger in size are closer to the surface of the device, while 495
dots that are smaller are buried farther down in the device. 496
At this voltage, we are above the threshold voltage for the 497
device; the channel is now heavily populated with carriers. 498
In the source and drain, the location of the electron density 499
depends rather weakly on the locations of the dopants. This 500
is mainly due to the scarcity of the dopants and that they are 501
buried deeply in the InAs substrate which reduces their effect 502
on the propagating electrons. Nevertheless, the density tends to 503
roughly follow the path of the dopants as it makes its way to the 504
channel. In the channel, we find that the some of the electrons 505
have been trapped by multiple sequential reflections off of the 506

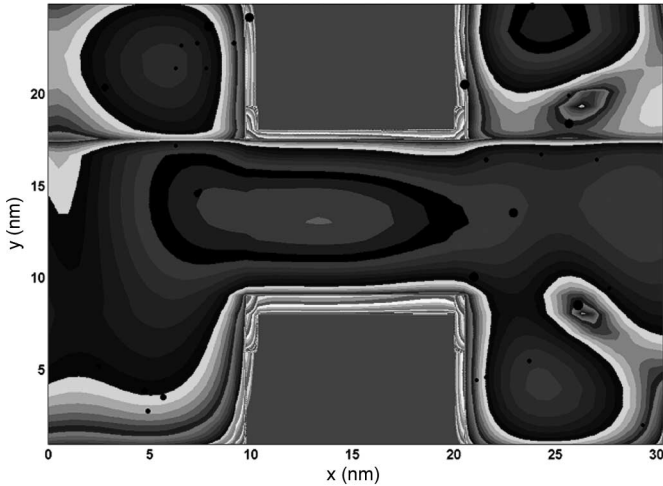


Fig. 11. Plot of the electron density in the xy plane taken at a depth of 5 nm into the InAs semiconductor layer at $V_g = V_d = 0.6$ V. The black dots represent the locations of the dopant atoms in the source and drain of the device. Larger dots represent atoms that are closer to the surface while smaller dots are more deeply buried in the semiconductor layer. At these biases, the channel is fully populated and the quantum interference is minimal.

507 source-channel and channel-drain interfaces. As the electrons
508 then exit the channel, they begin to populate sites in the drain
509 that are energetically preferential.

510 In Fig. 12, we plot the density at $V_g = 0.6$ V for the combined
511 (elastic + inelastic) case. We see very distinct differences in
512 the location and magnitude of the electrons in the combined
513 case, but, in general, it follows the same general trend as in the
514 elastic case. With the combined case, we find that there is a
515 pronounced buildup of charge in the area of the source near the
516 channel entrance of $0.38 k_b T$. The reflection off of this interface
517 alters the energy of the carriers to the point where some of
518 the lower energy modes will see increased scattering and a
519 reduction in density of approximately $0.25 k_b T$ in most areas.
520 In the source, we find additional discrepancies as this is the
521 location where the scattering is most prevalent. In the channel
522 and drain of the device, we find that there is a corresponding
523 reduction of electron density as much of the incident density
524 has been reflected at the channel interface.

525

V. CONCLUSION

526 In this paper, we have presented a method for including
527 the nonlocal effects of POP scattering into simulation tech-
528 niques relying on a propagating Hamiltonian. We then used
529 this method to simulate III–V NWTs while including the
530 effects of not only POP scattering, but also impurity, intervalley,
531 and acoustic phonon scattering. In InAs devices, we find that
532 NWTs have a significant ballistic length at room temperature
533 of about 12.33 nm. This ballistic length increases nonlinearly
534 as the drain voltage increases as the nonlocal POPs become
535 localized at higher energy. Further, we find that reductions in
536 temperature will not improve the ballistic-to-diffusive crossover
537 length, but cooling the device to around 100 K improves the
538 performance greatly due to weak phonon interactions. InSb
539 devices have similar device performance, with a ballistic-
540 to-diffusive crossover of about 8.7 nm. The great difference

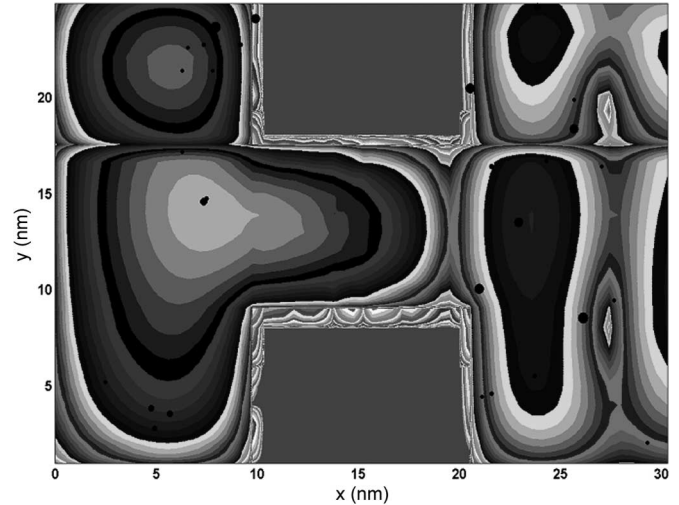


Fig. 12. Plot of the electron density in the xy plane taken at a depth of 5 nm into the InAs semiconductor layer at $V_g = V_d = 0.6$ V. The black dots represent the locations of the dopant atoms in the source and drain of the device. Here, there is significant charge build-up near the source-channel interface due to the combination of interference with the dopants near this interface and POPs.

between InAs and InSb is the locations of longitudinal states
541 in the channel which shifts a highly resistive region responsible
542 for poor device performance in both materials. This is not
543 the case with InP where we find that the POP scattering is
544 too strong and diffusive transport is achieved in the source of
545 the device. We simulated several 10-nm InAs devices to find
546 that at high gate voltages they behave as ballistic devices, but
547 the placement of dopants in the source and drain is crucial
548 to device performance. These results demonstrate the potential
549 and limitations of III–V NWTs for use in highly scaled CMOS
550 architectures. With ballistic lengths nine times longer than
551 silicon, they are certainly worthy of continued exploration for
552 novel electron quantum interference and waveguide devices. 553

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REFERENCES

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- [1] B. S. Doyle, S. Datta, M. Doczy, S. Hareland, B. Jin, J. Kavalieros, T. Linton, A. Murthy, R. Rios, and R. Chau, "High performance fully-depleted tri-gate CMOS transistors," *IEEE Electron Device Lett.*, vol. 24, no. 4, pp. 463–465, Apr. 2003. 558
- [2] T. Ashley, A. B. Dean, C. T. Elliott, G. J. Pryce, A. D. Johnson, and H. Willis, "Uncooled high-speed InSb field-effect transistors," *Appl. Phys. Lett.*, vol. 66, no. 4, pp. 481–483, Jan. 1995. 562
- [3] T. Bryllert, L.-E. Wernersson, L. E. Fröberg, and L. Samuelson, "Vertical high-mobility wrap-gated InAs nanowire transistor," *IEEE Electron Device Lett.*, vol. 27, no. 5, pp. 323–325, May 2006. 566
- [4] M. J. Gilbert and D. K. Ferry, "Efficient quantum three-dimensional modeling of fully-depleted ballistic silicon-on-insulator metal-oxide-semiconductor field-effect transistors," *J. Appl. Phys.*, vol. 95, no. 12, pp. 7954–7960, Jun. 2004. 570
- [5] —, "Quantum interference in fully-depleted tri-gate quantum wire transistors—The role of inelastic scattering," *IEEE Trans. Nanotechnol.*, vol. 4, no. 5, pp. 599–604, Sep. 2005. 574
- [6] J. Wang, E. Polizzi, and M. Lundstrom, "A three-dimensional quantum simulation of silicon nanowire transistors with the effective-mass approximation," *J. Appl. Phys.*, vol. 96, no. 4, pp. 2192–2203, Aug. 2004. 577

- 578 [7] M. J. Gilbert and D. K. Ferry, "Indium arsenide quantum wire trigate
579 metal oxide semiconductor field effect transistor," *J. Appl. Phys.*, vol. 99,
580 no. 12, p. 054503, 2006.
- 581 [8] M. J. Gilbert, R. Akis, and D. K. Ferry, "Phonon assisted ballistic to
582 diffusive crossover in silicon nanowire transistors," *J. Appl. Phys.*, vol. 98,
583 no. 9, p. 094303, Nov. 2005.
- 584 [9] M. J. Gilbert and D. K. Ferry, "Discrete dopant effects in ultra-small fully
585 depleted ballistic SOI MOSFETs," *Superlattices Microstruct.*, vol. 34,
586 no. 3–6, pp. 277–282, Sep. 2003.
- 587 [10] D. D. Johnson, "Modified Broyden's method for accelerating convergence
588 in self-consistent calculations," *Phys. Rev. B, Condens. Matter*, vol. 38,
589 no. 18, pp. 12807–12813, Dec. 1988.
- 590 [11] Y. Wang, J. Wang, H. Guo, and E. Zaremba, "Many electron effects
591 on ballistic transport," *Phys. Rev. B, Condens. Matter*, vol. 52, no. 4,
592 pp. 2738–2746, Jul. 1995.
- 593 [12] V. A. Sablikov, S. V. Polyakov, and M. Büttiker, "Charging effects in a
594 quantum wire with leads," *Phys. Rev. B, Condens. Matter*, vol. 61, no. 20,
595 pp. 13763–13773, May 2000.
- 596 [13] G. Kirczenow, "Resonant conduction in ballistic quantum channels,"
597 *Phys. Rev. B, Condens. Matter*, vol. 39, no. 14, pp. 10452–10455,
598 May 1989.
- 599 [14] A. T. Tilke, F. C. Simmel, H. Lorenz, R. H. Blick, and J. P. Kotthaus,
600 "Quantum interference in a one-dimensional silicon nanowire," *Phys. Rev.*
601 *B, Condens. Matter*, vol. 68, no. 7, p. 075311, Aug. 2003.



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